

Micropower Quad Comparator

FEATURES

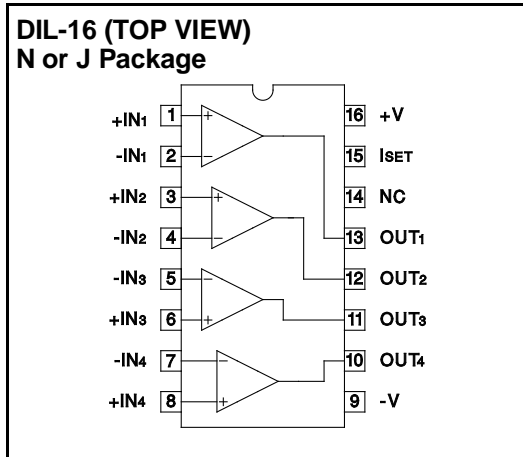
- Programmable Output Drive Capability
- Direct CMOS Logic Compatibility
- Low Power
- Direct Wire-OR of Outputs
- Wide Input Common Mode Range

DESCRIPTION

The UC161 family of quad comparators feature programmable DC and AC parameters. A single external resistor can set the comparators to operate in the microwatt region for battery applications, or higher current levels can be set to obtain improved speed or drive capabilities. The outputs on these devices can be wire OR'd together, simplifying external logic requirements in some applications.

These devices are available in three temperature ranges, the UC161A is specified for the full military range, -55° C to +125°C, the UC161B for the industrial range, -25°C to +85°C, and the UC161C for the commercial range of 0°C to +70°C.

CONNECTION DIAGRAM

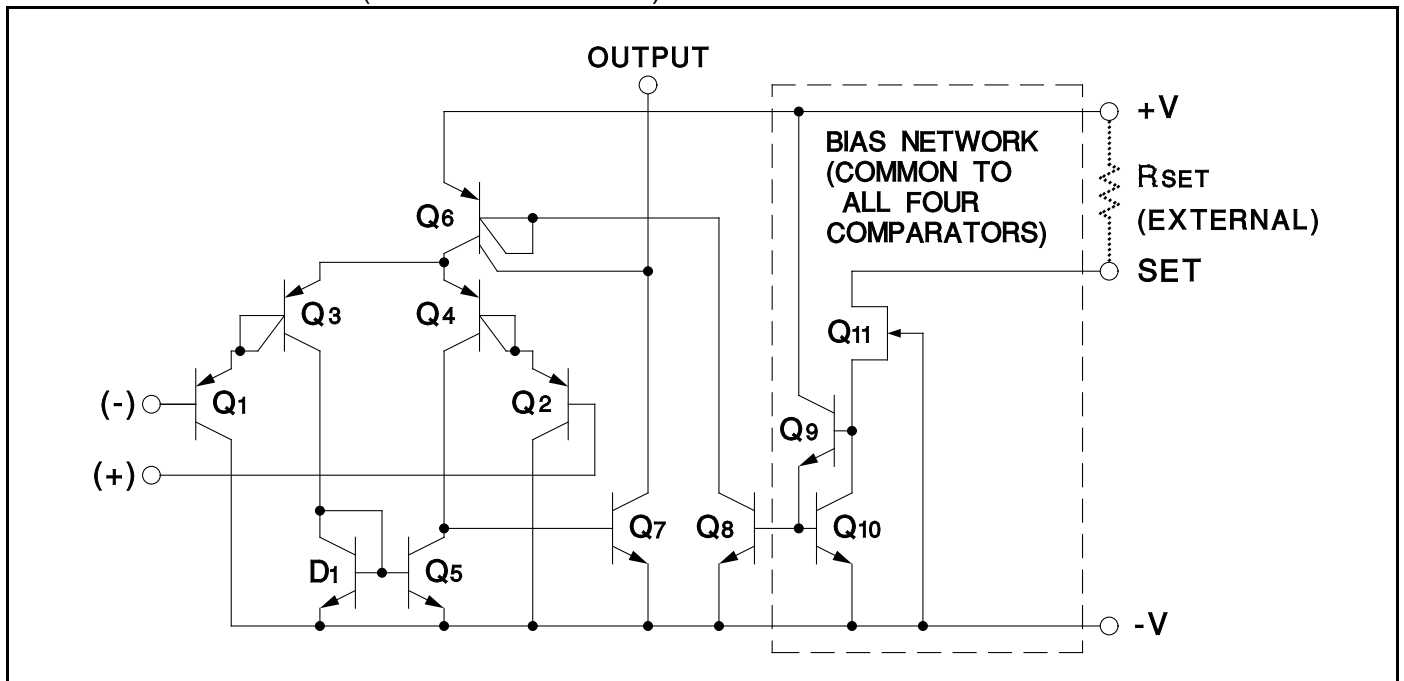


ABSOLUTE MAXIMUM RATINGS

| | |
|---------------------------------------|-----------------|
| Supply Voltage (+V to -V) | 36V |
| Differential Input Voltage | ±30V |
| Input Voltage | -V-0.3V to +V |
| Power Dissipation at TA = 25°C | 1000 mW |
| Power Dissipation at Tc = 25°C | 2000 mW |
| Operating Junction Temperature | -55°C to +150°C |
| Storage Temperature | -65°C to +150°C |
| Lead Temperature (Soldering, 10 Sec.) | +300°C |

Note: Consult Packaging Section of Databook for thermal limitations and considerations of package.

SIMPLIFIED SCHEMATIC (ONE COMPARATOR)



ELECTRICAL CHARACTERISTICS: Temperature range is -55° to +125°C for the UC161A, -25°C to +85°C for the UC161B, and 0°C to +70°C for the UC161C.

LOW POWER ELECTRICAL CHARACTERISTICS: Unless Otherwise Stated: $V_s = \pm 3V$, $I_{SET}^2 = 10\mu A$, $R_2 = 10M\Omega$, $C_L = 10pF$, $T_A = 25^\circ C$, $T_A = T_J$.

| | PARAMETER | SYMBOL | TEST CONDITIONS | UC161A | | | UC161B/C | | | UNITS |
|---|----------------------------------|------------------|---|--------|---------|------|----------|-------|------|-------|
| | | | | MIN | TYP | MAX | MIN | TYP | MAX | |
| INPUT | Input Offset Voltage | V _{OS} | | | 1 | 3 | | 1 | 6 | mV |
| | Input Offset Current | I _{OS} | | | 1 | 20 | | 1 | 25 | nA |
| | Input Bias Current | I _{BT} | | | 20 | 100 | | 20 | 200 | nA |
| OUTPUT | DC Open Loop Voltage Gain | A _{VOL} | | 20 | 30 | | 10 | 30 | | V/mV |
| | Low Output Voltage ¹ | V _{OL} | R _L = 20kΩ | | -2.95 | -2.6 | | -2.95 | -2.6 | V |
| | High Output Voltage ¹ | V _{OH} | R _L = 200kΩ | 2.5 | 2.9 | | 2.5 | 2.9 | | V |
| DYNAMIC | Common Mode Range | CMR | | | +1.3/-3 | | +1.3/-3 | | | V |
| | Response Time | t | 100mV Overdrive, C _L = 10pF | | 5 | | 5 | | | μs |
| | Common Mode Rejection Ratio | CMRR | V _{IN} = CMR | 75 | 90 | | 75 | 90 | | dB |
| SUPPLY | Power Supply Rejection Ratio | PSRR | | 65 | 80 | | 65 | 80 | | dB |
| | Supply Current | I _S | All Inputs Grounded, R _L = ∞ | | 210 | 300 | | 210 | 300 | μA |
| T_A = Over Temperature Range | | | | | | | | | | |
| | Input Offset Voltage | V _{OS} | | | | 5 | | | | mV |
| | DC Open Loop Voltage Gain | A _{VOL} | | 10 | | | 5 | | | V/mV |
| | Supply Current | I _S | All Inputs Grounded, R _L = ∞ | | | 350 | | | 350 | μA |

HIGH POWER ELECTRICAL CHARACTERISTICS: Unless Otherwise Stated: $V_s = \pm 15V$, $I_{SET}^2 = 100\mu A$, $R_L = 2M\Omega$, $C_L = 10pF$, $T_A = 25^\circ C$, $T_A = T_J$.

| | PARAMETER | SYMBOL | TEST CONDITIONS | UC161A | | | UC161B/C | | | UNITS |
|---------|----------------------------------|------------------|---|--------|---------|-------|----------|-------|-------|-------|
| | | | | MIN | TYP | MAX | MIN | TYP | MAX | |
| INPUT | Input Offset Voltage | V _{OS} | | | 1.5 | 3 | | 1.5 | 6 | mV |
| | Input Offset Current | I _{OS} | | | 5 | 60 | | 5 | 90 | nA |
| | Input Bias Current | I _{BT} | | | 100 | 400 | | 100 | 800 | nA |
| OUTPUT | DC Open Loop Voltage Gain | A _{VOL} | | 50 | 100 | | 30 | 100 | | V/mV |
| | Low Output Voltage ¹ | V _{OL} | R _L = 20kΩ | | -14.9 | -14.6 | | -14.9 | -14.6 | V |
| | High Output Voltage ¹ | V _{OH} | R _L = 200kΩ | 14.5 | 14.9 | | 14.5 | 14.9 | | V |
| DYNAMIC | Common Mode Range | CMR | | | +13/-15 | | +13/-15 | | | V |
| | Response Time | t | 100mV Overdrive, C _L = 10pF | | 1 | | 1 | | | μs |
| | Common Mode Rejection Ratio | CMRR | V _{IN} = CMR | 75 | 90 | | 75 | 90 | | dB |
| SUPPLY | Power Supply Rejection Ratio | PSRR | | 65 | 80 | | 65 | 80 | | dB |
| | Supply Current | I _S | All Inputs Grounded, R _L = ∞ | | 2100 | 3500 | | 2100 | 3500 | μA |

Note 1: The output current drive of the UC161 is non-symmetrical. This facilitates the wire-ORing of two comparator outputs. The output pull-down current capability is typically 75–150 times the pull-up current.

Note 2: Set current (I_{SET}) and supply current (I_{SUPPLY}) can be determined by the following formulas:

$$I_{SET} = \frac{[(+V) - (2V_{BE}) - (-V)]}{R_{SET}} ; I_{SUPPLY} = 21 \times I_{SET}$$

HIGH POWER ELECTRICAL CHARACTERISTICS (Continued): $T_A = T_J$.

$T_A =$ Over Temperature Range

| PARAMETER | SYMBOL | TEST CONDITIONS | UC161A | | | UC161B/C | | | UNITS |
|---------------------------|------------------|--|--------|-----|------|----------|-----|------|-------|
| | | | MIN | TYP | MAX | MIN | TYP | MAX | |
| Input Offset Voltage | V _{OS} | | | | 6 | | | | mV |
| Input Bias Current | I _{BT} | | | | 500 | | | | nA |
| DC Open Loop Voltage Gain | A _{VOL} | | 25 | | | 15 | | | V/mV |
| Supply Current | I _S | All Inputs Grounded R _L = ∞ | | | 4000 | | | 4000 | μA |

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APPLICATION AND OPERATION INFORMATION DESCRIPTION

The UC161 is a monolithic quad micropower comparator with an external control for varying its AC and DC characteristics. The variation of a single programming resistor will simultaneously alter parameters such as supply current, input bias, current slew rate, output drive capability, and gain. By making this resistor large, operation at very small supply current levels and power dissipations is possible. The UC161 is therefore ideal for systems requiring minimum power drain, such as battery-powered instrumentation, aerospace systems, CMOS designs, and remote security systems.

The circuit (see Simplified Schematic) is composed of five major blocks—four comparators and a common bias network. Q₁-Q₆, and D₁ from a darlington differential amplifier with double-to-single ended conversion. Q₆ is a dual current source whose outputs are exactly twice the current flowing through Q₈. The collector current of Q₈ is a function of the current supplied externally to Q₉-Q₁₀, which in turn is known as the set current of I_{SET}. This set current is established by a resistor connected between the I_{SET} terminal and a voltage source, most commonly the positive supply. Q₁₁ prevents excessive current from flowing through Q₉ and Q₁₀ in the event the I_{SET} terminal is shorted to the positive supply; it has no effect on circuit operation under normal conditions.

SETTING THE SET CURRENT

The set current can be expressed as:

$$I_{SET} = \frac{[(+V) - (2V_{BE}) - (-V)]}{R_{SET}}$$

where +V is the voltage to which the control resistor is connected, -V is the negative supply voltage, V_{BE} is the base emitter drop of Q₉ or Q₁₀ (about 0.7V), and R_{SET} is the value of the external control resistor or set resistor. Equation 1 is simply a derivative of ohms law. There is also an analytical relationship between I_{SET} and the total supply current:

$$\begin{aligned} I_{SUPPLY} &= [I_{SET} \text{ (current sourced by } Q_6 \text{ to } Q_8) \\ &\quad + 2 I_{SET} \text{ (current sourced to the differential amplifier by } Q_6) \\ &\quad + 2 I_{SET} \text{ (current sourced to the comparator output by } Q_6) \\ &\quad \times 4 \text{ (the total number of comparators)} \\ &\quad + I_{SET} \text{ (current sourced through } Q_{11}, Q_{10}, \text{ and } Q_9 \text{ to } -V)] \\ &= [I_{SET} + 2 I_{SET} + 2 I_{SET}] \times 4 + I_{SET} \\ &= 21 I_{SET} \end{aligned}$$

The output current pulldown capability (I_{OL}) of the UC161 is about 2 orders of magnitude greater than the high output drive current, (I_{OH}), which allows wire-ORing the outputs. I_{OH} is simply the current sourced by Q₆:

$$I_{OH} = 2 \times I_{SET}$$

I_{OL} is found by multiplying the current sourced by the collector of Q₆ by the gain Q₇:

$$I_{OL} = \beta (Q_7) \times 2 I_{SET}$$

The beta of Q₇ is about 75–150.

APPLICATION AND OPERATION INFORMATION (Continued)

